Exact Solution for Simply Supported and Multilayered Magneto-Electro-Elastic Plates

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Introduction

Because of their analytical nature, exact solutions for simply-supported (layered) plates under static loadings are still of particular values. These solutions can predict exactly the behaviors of elastic deformations and stresses near or across the interface of material layers, and can thus be used to check the accuracy of various numerical methods for more complicated applications ([1]). For anisotropic elastic composites, Pagano [2,3], Srinivas et al. [4], and Srinivas and Rao [5] derived the classic solutions for both the cylindrical and rectangular plates. While the author ([6]) introduced the propagator matrix method ([7]) to handle the corresponding multilayered case, Noor and Burton [8] derived analytical solutions for laminated anisotropic plates.

Recent development of piezoelectric ceramics has stimulated considerable studies on the electric and mechanical behaviors of piezoelectric structures. Again, analytical solutions, even though under certain assumptions, are still desirable. Extensions of the elastostatic solutions for simply-supported plates to the corresponding piezoelectric cases were carried out by Ray and co-workers [9,10], Heyliger and co-workers [11,12], Bisegna and Maceri [13], and Lee and Jiang [14]. Very recently, Vel and Batra [15] presented an analytical solution for multilayered piezoelectric plates in terms of the double Fourier series to handle more general boundary conditions at the edges.

More recent advances are the smart or intelligent materials among magnetic, electric, and mechanical energies. These new numerical results should be of special interest to the design of magneto-electro-elastic composite laminates.

Exact solutions are derived for three-dimensional, anisotropic, linearly magneto-electro-elastic, simply-supported, and multilayered rectangular plates under static loadings. While the homogeneous solutions are obtained in terms of a new and simple formalism that resembles the Stroh formalism, solutions for multilayered plates are expressed in terms of the propagator matrix. The present solutions include all the previous solutions, such as piezoelectric, piezomagnetic, purely elastic solutions, as special cases, and can therefore serve as benchmarks to check various thick plate theories and numerical methods used for the modeling of layered composite structures. Typical numerical examples are presented and discussed for layered piezoelectric/piezomagnetic plates under surface and internal loads. [DOI: 10.1115/1.1380385]

Problem Description and Basic Equations

Let us consider an anisotropic, magneto-electro-elastic, and N-layered rectangular plate with horizontal dimensions Lx and Ly, and total thickness H (in the vertical direction) with its four sides being simply supported. A Cartesian coordinate system \((x, y, z) = (x_1, x_2, x_3)\) is attached to the plate in such a way that its origin is at one of the four corners on the bottom surface and the plate is in the positive z region. Let layer \(j\) be bonded by the lower interface \(z_j\) and the upper interface \(z_{j+1}\) with thickness \(h_j = z_{j+1} - z_j\). It is obvious that \(z_1 = 0\) and \(z_{N+1} = H\). Material properties in each layer can be different, and internal and/or surface loads (mechanical, electric or magnetic) can be applied. Along the interface, the extended displacement and traction vectors (to be defined later) are assumed to be continuous, with the exception of the internal loading level, which will be discussed later. Without loss of generality, we also assume that the surface load is applied on the top surface of the layered plate.
For an anisotropic and linearly magneto-electro-elastic solid, the coupled constitutive relation can be written as ([16])
\[
\sigma_i = C_{ik} \gamma_k - e_{ik} E_k - q_{ik} H_k
\]
\[
D_i = e_{ij} \gamma_j + i e_{ij} E_j + \mu_{ij} H_j
\]
\[
B_i = q_{ik} \gamma_j + d_{ik} E_j + \mu_{ik} H_k
\]
where \(\sigma_i\), \(D_i\), and \(B_i\) are the stress, electric displacement, and magnetic induction (i.e., magnetic flux), respectively; \(\gamma_i\), \(E_i\), and \(H_i\) are the strain, electric field, and magnetic field, respectively; \(C_{ij}\), \(e_{ij}\), and \(\mu_{ij}\) are the elastic, dielectric, and magnetic permeability coefficients, respectively; \(e_{ij}\), \(q_{ij}\), and \(d_{ij}\) are the piezoelectric, piezomagnetic, and magnetoelectric coefficients, respectively. It is obvious that various uncoupled cases can be reduced from Eq. (1) by setting the appropriate coefficients to zero.

For an orthotropic solid, with transverse isotropy being a special case, the material constant matrices of Eq. (1) are expressed by
\[
[C] = \begin{bmatrix}
C_{11} & C_{12} & C_{13} & 0 & 0 & 0 \\
C_{22} & C_{23} & 0 & 0 & 0 \\
C_{33} & 0 & 0 & 0 & 0 \\
C_{44} & 0 & 0 & 0 & 0 \\
\text{Sym} & C_{55} & 0 & 0 & 0 \\
C_{66} & & & & \\
\end{bmatrix}
\]
\[
[e] = \begin{bmatrix}
e_{11} & 0 & 0 & 0 \\
e_{22} & 0 & 0 & 0 \\
e_{33} & 0 & 0 & 0 \\
e_{23} & 0 & 0 & 0 \\
e_{13} & 0 & 0 & 0 \\
e_{12} & 0 & 0 & 0 \\
\end{bmatrix}
\]
\[
[q] = \begin{bmatrix}
q_{11} & 0 & 0 & 0 \\
q_{22} & 0 & 0 & 0 \\
q_{33} & 0 & 0 & 0 \\
q_{23} & 0 & 0 & 0 \\
q_{13} & 0 & 0 & 0 \\
q_{12} & 0 & 0 & 0 \\
\end{bmatrix}
\]
\[
[d] = \begin{bmatrix}
d_{11} & 0 & 0 & 0 \\
d_{22} & 0 & 0 & 0 \\
d_{33} & 0 & 0 & 0 \\
\end{bmatrix}
\]
\[
[\mu] = \begin{bmatrix}
\mu_{11} & 0 & 0 & 0 \\
0 & \mu_{22} & 0 & 0 \\
0 & 0 & \mu_{33} & 0 \\
\end{bmatrix}
\]

The extended strain (using tensor symbol for the elastic strain \(\gamma_{ik}\)-displacement relation is
\[
\gamma_{ij} = 0.5(u_{ij} + u_{ji})
\]
\[
E_i = -\phi_i, \quad H_i = -\psi_i
\]
where \(u_i\), \(\phi\), and \(\psi\) are the elastic displacement, electric potential, and magnetic potential, respectively.

The equations of equilibrium, including the balance of the body force and electric charge and current, can be written as
\[
\sigma_{ij,i} + f_i = 0
\]
\[
D_{ij,j} - f_e = 0
\]
\[
B_{ij,j} - f_m = 0
\]
where \(f_i\), \(f_e\), and \(f_m\) are the body force, electric charge density, and electric current density, respectively. The electric current density is also called magnetic charge density as compared to the electric charge density.

\[\text{General Solutions}\]
For a simply-supported and homogeneous plate, we seek the solution of the extended displacement vector in the form of
\[
u = \begin{bmatrix} u_1 \\ u_2 \\ \phi \\ \psi \end{bmatrix} = e^{i\omega t} \begin{bmatrix} a_1 \\ a_2 \\ a_3 \\ a_4 \end{bmatrix}
\]
where \(p = n \pi/L_x\), \(q = m \pi/L_y\)

Introducing two vectors
\[
a = [a_1, a_2, a_3, a_4, a_5]^T, \quad \text{b} = [b_1, b_2, b_3, b_4, b_5]^T
\]
we then find that the vector \(\text{b}\) is related to \(\text{a}\) by
\[
\text{b} = (-R + sT)a = -\frac{1}{s}(Q + sR)a
\]
where the superscript \(t\) denotes matrix transpose, and
\[
R = \begin{bmatrix}
0 & 0 & pC_{13} & pe_{31} & pq_{31} \\
0 & 0 & qC_{23} & qe_{32} & pq_{32} \\
-pC_{55} & -qC_{44} & 0 & 0 & 0 \\
-p e_{15} & -q e_{24} & 0 & 0 & 0 \\
-pq_{15} & -q q_{24} & 0 & 0 & 0 \\
\end{bmatrix}
\]
\[
T = \begin{bmatrix}
C_{55} & 0 & 0 & 0 & 0 \\
C_{44} & 0 & 0 & 0 & 0 \\
C_{33} & e_{33} & q_{33} & -e_{33} & -d_{33} \\
-\mu_{33} & & & &
\end{bmatrix}
\]
We mention that matrices $Q$ and $T$ are symmetric. The in-plane stresses and electric and magnetic displacements are obtained as

$$\sigma_{11} = \sigma_{22} = \sigma_{ij} = \frac{1}{2}(\sigma_{ij} + \sigma_{ji}), \quad \varepsilon_{11} = \varepsilon_{22} = \varepsilon_{ij} = \frac{1}{2}(\sigma_{ij} - \sigma_{ji}),$$

where

$$\begin{bmatrix}
-\frac{C_{11}}{2} & \frac{C_{12}}{2} & \frac{C_{13}}{2} & \frac{e_{31}}{2} & \frac{q_{31}}{2} \\
\frac{C_{12}}{2} & \frac{C_{22}}{2} & \frac{C_{23}}{2} & \frac{e_{32}}{2} & \frac{q_{32}}{2} \\
\frac{C_{13}}{2} & \frac{C_{23}}{2} & \frac{C_{33}}{2} & \frac{e_{33}}{2} & \frac{q_{33}}{2} \\
\frac{e_{31}}{2} & \frac{e_{32}}{2} & \frac{e_{33}}{2} & \frac{a_{1}}{2} & \frac{a_{2}}{2} \\
\frac{q_{31}}{2} & \frac{q_{32}}{2} & \frac{q_{33}}{2} & \frac{a_{3}}{2} & \frac{a_{4}}{2} \\
\end{bmatrix}$$

(13)

These extended stresses (Eqs. (8) and (13)) should satisfy the equations of equilibrium (assuming zero body force and zero electric and magnetic charge densities), which in terms of the vector $a$, yields the following eigenequation:

$$[Q + s(R + R^T) + s^2T]a = 0$$

(15)

where $R' = -R$. It is noted that Eq. (15), derived for a simply supported plate, resembles the Stroh formalism ([25,26]). However, their solution structures are different because of the slightly different features of the $R$ matrix (in the Stroh formalism, $R' = R$). It is known that in the Stroh formalism, positive internal energy requirement guarantees that the characteristic roots of Eq. (15) should be complex numbers with nonvanishing imaginary parts; they cannot be real ([26]). In the present formalism, however, such a feature does not exist. Instead, since a matrix and its transpose have the same determinant value, we conclude that if $s$ is an eigenvalue of Eq. (15), so is $-s$. Furthermore, if $s$ is a complex (or purely imaginary) eigenvalue, then its complex conjugate is also an eigenvalue since all the coefficients matrices in Eq. (15) are real. We name Eq. (15) as the pseudo-Stroh formalism because of its similarity to the Stroh formalism.

With aid of Eq. (10), Eq. (15) can now be recast into a $10 \times 10$ linear eigensystem

$$N[a] = [a]$$

(16)

where

$$N = \begin{bmatrix}
-T^{-1}R' & T^{-1} \\
-Q + RT^{-1}R' & -RT^{-1} \\
\end{bmatrix}$$

(17)

Depending upon the given material property, the ten eigenvalues of Eq. (16) may not be distinct. Should repeated roots occur, a slight change in the material constants would result in distinct roots with negligible error ([28]) so that the following simple solution structure can still be applied.

Therefore, let us assume that the first five eigenvalues have positive real parts (if the root is purely imaginary, we then pick up the one with positive imaginary part) and the remainder have opposite signs to the first five. We distinguish the corresponding ten eigenvectors by attaching a subscript to $a$ and $b$. Then the general solution for the extended displacement and traction vectors (of the $z$-dependent factor) are derived as

$$\begin{bmatrix}
u \\ t\end{bmatrix} = \begin{bmatrix} A_1 & A_2 \end{bmatrix} (e^{s_1z})K_1$$

(18)

where

$$A_1 = [a_1,a_2,a_3,a_4,a_5], \quad A_2 = [a_6,a_7,a_8,a_9,a_{10}]$$

and $K_1$ and $K_2$ are two $5 \times 1$ constant column matrices to be determined.

Equation (18) is a general solution for a homogeneous, magneto-electro-elastic, and simply-supported plate, and contains previous piezoelectric and purely elastic solutions as its special cases. Clearly, in spite of the complicated nature of the problem, the general solution is remarkably simple. Furthermore, certain thin plate results can also be reduced from this solution by expanding the exponential term in terms of a Taylor series ([29,30]). This is particularly easy since one needs only to replace the diagonal exponential matrix with its Taylor series expansion ([6,13]). We mention that although other methods, such as the state space approach ([14]), may also be employed to derive a general solution for such a plate, more algebraic manipulations are needed. Furthermore, reduction to the thin plate result is complicated if a state space approach is followed.

With Eq. (18) being served as a general solution for a homogeneous and magneto-electro-elastic plate, solutions for the corresponding multilayered plate can be obtained using the continuity conditions along the interface and the boundary conditions on the top and bottom surfaces of the plate. In doing so, a system of linear equations for the unknowns can be formed and solved ([3,12]). However, for structures with relatively large number of layers (say, up to a hundred layers), the system of linear equations then becomes very large, and the propagator matrix method developed exclusively for layered structures can be conveniently and efficiently applied (for a brief review, see [31]). We discuss this matter in the next section.

**Propagator Matrix and Solution of Layered System**

Since the matrix $N$, defined in Eq. (17), is not symmetric, the eigenvectors of Eq. (16) are actually the right ones. The left eigenvectors are found by solving the following eigenvalue system:

$$N' \eta = \lambda \eta$$

(19)
It is a matter of simple fact that if \( s \) and \([a, b]^T\) are the eigenvalue and eigenvector of Eq. (16), then \( \lambda = -s \) and \( \eta = [\ -b, a]^T\) are the corresponding solutions of Eq. (19). Since the left and right eigenvectors are orthogonal to each other, we then come to the following important relation:

\[
\begin{bmatrix}
B_2 & A_2 \\
B_1 & A_1
\end{bmatrix} \begin{bmatrix}
A_1 & A_2 \\
B_1 & B_2
\end{bmatrix} = \begin{bmatrix}
I & 0 \\
0 & I
\end{bmatrix}
\]

(20)

where \( I \) is a \( 5 \times 5 \) unit matrix, and the eigenvectors have been normalized according to

\[
-b_2 A_1 + a_1 B_1 = I
\]

(21)

Equation (20) resembles the orthogonal relation in the Stroh formalism ([26]) and provides us with a simple way of inverting the eigenvector matrix, which is required in forming the propagator matrix.

Let us assume that Eq. (18) is a general solution in the homogeneous layer \( j \), with the top and bottom boundaries locally at \( h \) and \( 0 \), respectively. Let \( z = 0 \) in Eq. (18) and solve for the unknown constant column matrix, we find that

\[
\mathbf{K}_j = \begin{bmatrix}
A_1 & A_2 \\
B_1 & B_2
\end{bmatrix}^{-1} \begin{bmatrix}
u(t) \\
0
\end{bmatrix} = \begin{bmatrix}
-B_2 & A_2 \\
B_1 & -A_1
\end{bmatrix} \begin{bmatrix}
u(t) \\
0
\end{bmatrix}.
\]

(22)

The second equation follows from Eq. (20). Therefore, the solution in the homogeneous layer \( j \) at any level \( z \) can be expressed by that at \( z = 0 \) as

\[
\begin{bmatrix}
u(t) \\
0
\end{bmatrix}_z = P(z) \begin{bmatrix}
u(t) \\
0
\end{bmatrix}_0
\]

(23)

where

\[
P(z) = \begin{bmatrix}
A_1 & A_2 \\
B_1 & B_2
\end{bmatrix} (e^{zA_1} - B_2) \begin{bmatrix}
-A_2 \\
B_1 - A_1
\end{bmatrix}
\]

(24)

is called the propagator matrix ([7,31]). Listed below are three important features of the propagator matrix, which can be proved easily.

\[
P(0) = \begin{bmatrix}
I & 0 \\
0 & I
\end{bmatrix}
\]

(25)

\[
P(z) = \begin{bmatrix}
A_1 & A_2 \\
B_1 & B_2
\end{bmatrix} P(z) \begin{bmatrix}
A_1 & A_2 \\
B_1 & B_2
\end{bmatrix}
\]

(26)

\[
P(z) = P \begin{bmatrix}
-A_2 \\
B_1 - A_1
\end{bmatrix} (z - z_1)
\]

(27)

The propagating relation (23) can be used repeatedly so that one can propagate the physical quantities from the bottom surface \( z = 0 \) to the top surface \( z = H \) of the layered plate. Consequently, we have

\[
\begin{bmatrix}
u(t) \\
0
\end{bmatrix}_j = P_j(h_j) P_{j-1}(h_{j-1}) \ldots \ldots P_2(h_2) P_1(h_1) \begin{bmatrix}
u(t) \\
0
\end{bmatrix}_0
\]

(28)

where \( h_j = z_{j+1} - z_j \) is the thickness of layer \( j \) and \( P_j \) the propagator matrix of the same layer.

Equation (28) is a surprisingly simple relation and, for given boundary conditions, can be solved for the unknowns involved. As an example, we assume that, on the top surface \( (z = H) \) the \( z \)-direction traction component is applied, i.e.,

\[
\sigma_{zz} = \sigma_0 \sin px \sin qy
\]

(29)

which may represent one of the terms in the double Fourier series solution for a general loading case (uniform or point loading), and all other traction components on both surfaces are zero (i.e., the second-type boundary value problem). Equation (28) is then reduced to

\[
\begin{bmatrix}
u(H) \\
t(H)
\end{bmatrix} = \begin{bmatrix}
C_1 & C_2 \\
C_3 & C_4
\end{bmatrix} \begin{bmatrix}
u(0) \\
0
\end{bmatrix}
\]

(30)

where the four submatrices \( C_j \) are the multiplications of the propagator matrices in Eq. (28), and \( t(H) \) is the given boundary condition on the top surface, i.e.,

\[
t(H) = [0,0, \sigma_0 \sin px \sin qy, 0,0]^T.
\]

(31)

Solving the unknown extended displacements on both surfaces of the layered plate, we find

\[
\begin{bmatrix}
u(0) \\
t(0)
\end{bmatrix} = C_1^{-1} t(H).
\]

(32)

In order to obtain the extended displacement and traction vectors at any depth, say \( z_k \leq z \leq z_{k+1} \) in layer \( k \), we propagate the solution from the bottom of the surface to the \( z \)-level ([31]), i.e.,

\[
\begin{bmatrix}
u(t) \\
t(t)
\end{bmatrix}_z = P_k(z - z_{k-1}) P_{k-1}(h_{k-1}) \ldots \ldots P_2(h_2) P_1(h_1) \begin{bmatrix}
u(t) \\
0
\end{bmatrix}_0.
\]

(33)

With the extended displacement and traction vectors at a given depth being solved, the corresponding in-plane quantities can be evaluated using Eqs. (13) and (14).

Similar solutions can also be obtained for the first-type boundary value problem (i.e., for given extended displacement vectors on both surfaces) and for the third-type, i.e., the mixed boundary value problem as well. Therefore, for an anisotropic, magneto-electro-elastic, and simply-supported multilayered rectangular plate, we have derived the exact solution based on the pseudo-Stroh formalism and the propagator matrix method.

The present methodology can also be equally and easily extended to the corresponding internal loading case, which is of significance to the Green’s function study. We now seek such a solution.

If there is an internal source (force, charge, dislocation, etc.) located at \( z = d_0 \) level within layer \( j(z_j, z_{j+1}) \), we artificially divide this layer into two sublayers \( j_1(d_0, z_j) \) (with \( h_{j_1} = d_0 - z_j \)) and \( j_2(z_{j+1}, d_0) \) (with \( h_{j_2} = z_{j+1} - d_0 \)), and define the discontinuities across the source level as

\[
\Delta u \quad \Delta t = \begin{bmatrix}
\nu(d_0 + 0) \\
t(d_0 + 0)
\end{bmatrix} - \begin{bmatrix}
\nu(d_0 - 0) \\
t(d_0 - 0)
\end{bmatrix}.
\]

(34)

Again, propagating the propagator matrices from the bottom to the top of the surfaces and making use of the discontinuity relation (34) ([31,32]), we arrive at the following important equation:

\[
\begin{bmatrix}
u(t) \\
0
\end{bmatrix}_H = P_n(h_n) P_{n-1}(h_{n-1}) \ldots \ldots P_2(h_2) P_1(h_1) \begin{bmatrix}
u(t) \\
0
\end{bmatrix}_0
\]

\[
= P_n(h_n) P_{n-1}(h_{n-1}) \ldots \ldots P_{j+1}(h_{j+1}) P_{j_2}(h_{j_2}) \begin{bmatrix}
\Delta u \\
\Delta t
\end{bmatrix}.
\]

(35)

Clearly, this equation is more general and includes Eq. (28) as a special case (when there is no discontinuity). Similar to the surface loading case, this equation can be solved for the unknown quantities involved ([31]).

Before carrying out numerical studies using the present formulation, we remark that the present solution is valid for any integers \( n \) and \( m \) as defined by Eq. (7). In other words, the solution we have derived can be regarded as for one of the terms in a Fourier series expansion. Because of the linearity, the solution corresponding to a general loading (uniform or point loading) can be obtained by expanding the loading as a finite double Fourier series ([13,33]) and adding the responses together term by term.

**Numerical Examples**

Having derived the exact and simple solutions, we now present some numerical results. Before using our formalism, we first checked our solutions with some previously published results for...
Table 1 Material coefficients of the piezoelectric BaTiO$_3$ ($C_{ij}$ in $10^9$ N/m$^2$, $e_{ij}$ in C/m$^2$, $\varepsilon_{ij}$ in $10^{-9}$ C/N m, and $\mu_{ij}$ in $10^{-6}$ Ns/C)

<table>
<thead>
<tr>
<th>$C_{11}$</th>
<th>$C_{12}$</th>
<th>$C_{13}$</th>
<th>$C_{33}$</th>
<th>$C_{44}$</th>
<th>$C_{66}$</th>
<th>$\varepsilon_{31}$</th>
<th>$\varepsilon_{32}$</th>
<th>$\varepsilon_{33}$</th>
<th>$\varepsilon_{24}$</th>
<th>$\varepsilon_{15}$</th>
<th>$\mu_{11}$</th>
<th>$\mu_{22}$</th>
<th>$\mu_{33}$</th>
</tr>
</thead>
<tbody>
<tr>
<td>166</td>
<td>77</td>
<td>78</td>
<td>162</td>
<td>43</td>
<td>43</td>
<td>-4.4</td>
<td>18.6</td>
<td>11.6</td>
<td>4.4</td>
<td>18.6</td>
<td>11.6</td>
<td>5</td>
<td>10</td>
</tr>
</tbody>
</table>

Fig. 1 Variation of the elastic displacement $u_x(=u_y)$ along the thickness direction in a homogeneous and piezoelectric plate caused by an internal load on the middle plane and a surface load on the top surface.

Fig. 2 Variation of the electric displacement $D_x(=D_y)$ along the thickness direction in a homogeneous and piezoelectric plate caused by an internal load on the middle plane and a surface load on the top surface.
Fig. 3 Variation of the stress component $\sigma_{zz}$ along the thickness direction in a homogeneous and piezoelectric plate caused by an internal load on the middle plane and a surface load on the top surface.

Table 2 Material coefficients of the magnetostrictive CoFe$_2$O$_4$ ($C_{ij}$ in $10^9$N/m$^2$, $q_{ij}$ in N/(Am), $\varepsilon_{ij}$ in $10^{-9}$C$^2$/Nm$^2$, and $\mu_{ij}$ in $10^{-6}$Ns$^2$/C$^2$)

<table>
<thead>
<tr>
<th>$C_{11}$ = $C_{22}$</th>
<th>$C_{12}$</th>
<th>$C_{13}$ = $C_{23}$</th>
<th>$C_{33}$</th>
<th>$C_{44}$ = $C_{55}$</th>
<th>$C_{66}$ = 0.5($C_{11}$ - $C_{12}$)</th>
</tr>
</thead>
<tbody>
<tr>
<td>286</td>
<td>173</td>
<td>170.5</td>
<td>269.5</td>
<td>45.3</td>
<td>56.5</td>
</tr>
</tbody>
</table>

$q_{31} = q_{32}$  
$q_{33}$  
$q_{24} = q_{15}$  
580.3  
699.7  
550

$\varepsilon_{11} = \varepsilon_{22}$  
$\varepsilon_{33}$  
$\mu_{11} = \mu_{22}$  
$\mu_{33}$  
0.08  
0.093  
-590  
157

Fig. 4 Variation of the electric potential $\phi$ along the thickness direction in the sandwich piezoelectric/piezomagnetic plate caused by a surface load on the top surface.
both purely elastic and piezoelectric plates ([3,12,14,34]), and found that the present formulation agrees with these solutions.

The first example is for a homogeneous and transversely isotropic piezoelectric plate. The symmetry axis of the material is along the $z$-direction with material properties being listed in Table 1 ([14]). The dimension of the plate is $L_x \times L_y \times H = 1 \times 1 \times 0.2$ m. Two cases are studied: (1) A $z$-direction surface load is applied on the top surface of the plate $z = H$. That is, the extended traction is given by Eq. (31) with $m = n = 1$ (i.e., $p = \pi/L_x$, $q = \pi/L_y$) and amplitude $\sigma_0 = 1$ N/m$^2$. The bottom surface is assumed to be traction-free. (2) An internal load is applied on the middle plane of the plate ($z = 0.1$ m). The extended traction discontinuity $\Delta t$ has a similar expression as Eq. (31) with amplitude $\Delta \sigma_{zz}$ equal to 1 N/m$^2$. Both the top and bottom surfaces are assumed to be traction-free. For both cases, responses are calculated for fixed horizontal coordinates $(x, y) = (0.75L_x, 0.25L_y)$.

Figures 1, 2, and 3 show the variations of the elastic displacement $u_x$, electric displacement $D_x$, and normal stress $\sigma_{zz}$ along

![Fig. 5 Variation of the magnetic potential $\psi$ along the thickness direction in the sandwich piezoelectric/piezomagnetic plate caused by a surface load on the top surface](image)

![Fig. 6 Variation of the electric displacement $D_x(=D_y)$ along the thickness direction in the sandwich piezoelectric/piezomagnetic plate caused by a surface load on the top surface](image)
the thickness direction of the plate. It is clear that these two loading cases produce quite different responses in the plate, even though the plate is relatively thin (with a ratio of thickness to horizontal dimension equal to 0.2). For instance, while the internal loading solution is strictly symmetric or antisymmetric with respect to the middle plane (i.e., the loading plane), the surface loading solution does not possess such features. The latter (for the elastic displacement $u_z$ and electric displacement $D_z$) is only approximately symmetric or antisymmetric about the middle plane. While the normal stress $\sigma_{zz}$ due to the surface load is continuous and increases monotonically from zero on the bottom surface to the applied value on the top surface, that due to the internal load is discontinuous across the loading plane $z=0.1$ m and it has opposite sign on both sides of the middle plane. The internal loading case has never been studied and compared to the surface loading case in the literature.

The second example is for sandwich plates made of piezoelectric BaTiO$_3$ and magnetostrictive CoFe$_2$O$_4$. The three layers have

![Fig. 7 Variation of the electric displacement $D_z$ along the thickness direction in the sandwich piezoelectric/piezomagnetic plate caused by a surface load on the top surface](image1)

![Fig. 8 Variation of the magnetic induction $B_x$ along the thickness direction in the sandwich piezoelectric/piezomagnetic plate caused by a surface load on the top surface](image2)
equal thickness of 0.1 m (with a total thickness \( H = 0.3 \) m). While the material properties for the piezoelectric BaTiO\(_3\) are those listed in Table 1, the properties for the magnetostrictive CoFe\(_2\)O\(_4\) are given in Table 2 ([35]). Similar to the piezoelectric BaTiO\(_3\), the magnetostrictive CoFe\(_2\)O\(_4\) is also a transversely isotropic solid with its symmetry axis along the \( z \)-axis.

Two sandwich plates with stacking sequences BaTiO\(_3\)/CoFe\(_2\)O\(_4\)/BaTiO\(_3\) (called B/F/B) and CoFe\(_2\)O\(_4\)/BaTiO\(_3\)/CoFe\(_2\)O\(_4\) (called F/B/F) are investigated. The surface loading as for the first example is assumed here (that is, a \( z \)-direction traction with amplitude \( \sigma_0 = 1 \) N/m\(^2\) is applied on the top surface \( z = 0.3 \) m while all other components on both surfaces are zero). Again, responses are calculated for fixed horizontal coordinates \((x,y) = (0.75L_x, 0.25L_y)\).

Figures 4 and 5 show, respectively, the variations of the electric and magnetic potentials along the thickness direction in the sandwich plate. It is obvious that the potential variations for the B/F/B
and F/B/F cases are completely different, demonstrating clearly the role played by the material stacking sequences. Furthermore, the slopes of these quantities can be discontinuous across the interface, even though the potentials themselves are continuous.

While Figs. 6 and 7 show the electric displacements \( D_y \) and \( D_z \), the magnetic displacements (magnetic induction) \( B_y \) and \( B_z \) are plotted in Figs. 8 and 9. The following general features are observed from these figures:

1. The horizontal electric and magnetic displacements are discontinuous across the interfaces (Figs. 6 and 8).

2. The magnitude of horizontal electric (magnetic) displacement is very small in magnetostrictive CoFe$_2$O$_4$ (piezoelectric BaTiO$_3$) layer (Figs. 6 and 8). This is due to the fact that for the magnetostrictive CoFe$_2$O$_4$ (piezoelectric BaTiO$_3$) material, the piezoelectric \( e_{ij} \) (piezomagnetic \( q_{ij} \)) coefficients are zero.

3. Within the outer layers, the horizontal and vertical electric displacements (magnetic inductions) change dramatically for the B/F/B (F/B/F) case (Figs. 6–9).

4. For these dramatically changed physical quantities, the vertical components reach their maximum magnitudes in the middle of the outer layers (Figs. 7 and 9), while for the horizontal components, the maxima are on the top and bottom surfaces and the minima at the interfaces.

The while electric and magnetic quantities have been greatly influenced by the stacking sequences, relatively small differences have been observed for the corresponding elastic displacements and stresses for these two sandwich cases. For instance, Fig. 10 shows the variation of the normal stress \( \sigma_z \) along the thickness direction in the sandwich piezoelectric/piezomagnetic plates. It is apparent that both stacking sequences produce nearly the same stress distribution, even though the elastic constants for the two materials are considerably different (Tables 1 and 2). This is obviously a coupling phenomenon and can only be explained by resorting to the coupled constitutive relation (1). For the stress field, it is seen from Eq. (1) that it consists of three parts: the elastic constant and strain, the piezoelectric coefficient and electric field, and the piezomagnetic coefficient and magnetic field. Even though the first part may produce quite different stresses in both sandwich plates, the effect of the second and third parts (i.e., the piezoelectric and piezomagnetic terms) is to wipe out, in the present case, the difference of the stress field produced by the first part.

The model results may have potential applications in the field of magneto/intelligent structures. For example, to design a sandwich plate made of the magnetostrictive CoFe$_2$O$_4$ and piezoelectric BaTiO$_3$ materials that requires a given stress level (or distribution) within the plate under a normal surface loading on the top, then in order to produce a large horizontal electric displacement \( D_y \) or \( D_z \) on both the top and bottom surfaces (Fig. 6), the B/F/B stacking sequence should be selected. On the other hand, if a large horizontal magnetic induction \( B_y \) or \( B_z \) on both the top and bottom surfaces (Fig. 8) is expected, then the F/B/F stacking sequence is the choice.

Conclusions

In this paper, we have derived exact solutions for three-dimensional, anisotropic magneto-electro-elastic, simply-supported, and multilayered rectangular plates under both surface and internal loads. We have developed a new and simple formalism that resembles the Stroh formalism so that the homogeneous solution can be obtained in a simple and elegant form. We have also introduced the propagator matrix method in order to treat efficiently and accurately the multilayered case. Our solutions include all the previous solutions, such as the piezoelectric, piezomagnetic, purely elastic solutions, as special cases, and can provide benchmarks for various thick plate theories and numerical methods, such as the finite and boundary element methods.

Two typical numerical examples presented have also shown some significant and interesting features. For instance, responses to an internal load are quite different from those to a surface load, even for a relatively thin plate. The solution to the internal load and its comparison to the corresponding surface loading solution have never been reported in the literature. For sandwich plates made of the piezoelectric BaTiO$_3$ and magnetostrictive CoFe$_2$O$_4$, we have observed that the stacking sequences (B/F/B and F/B/F) have a clear influence on most physical quantities, in particular, on the electric and magnetic quantities. These features should be of special interest to the design of magneto-electro-elastic composite laminates.

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References